

International Journal of Intelligent Engineering & Systems

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Performance of Max-Min User Relay Cooperative NOMA Systems with Imperfect CSI

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Abstract: In this paper, we derive the effect of channel errors on the performance of user relaying cooperative NOMA system in closed-form. The max-min selection strategy is adapted to the selection of a relay among near users (NUs) and selection combining (SC) is applied to a far user (FU) for the performance improvements. As the previous studies found, the max-min relay selection without channel errors is effective to improve the outage performance. However, we demonstrate that the outage performance is highly sensitive to the channel error. It is also found that the outage probability of the relay user is more susceptible to the channel errors. As the channel correlation coefficient approaches to 0, the performance gains are banishing and the outage probability with the max-min relay selection becomes identical to that with the random relay selection. NOMA system has an optimal power allocation coefficient to minimize the outage performance, but as the correlation coefficient decrease, the effect of power allocation to the performance becomes lessen. The provided numerical results confirm that the derived expressions match well with Monte Carlo simulations.

Keywords: NOMA, Cooperative NOMA, Max-min, Imperfect CSI, User relay.

1. Introduction

Next-generation cellular mobile systems demand more traffic capacity and higher data rates within the limited bandwidth. Recently introduced nonorthogonal multiple access (NOMA) has been focused for the promising candidate for the nextgeneration cellular networks. In NOMA, multiple users share both time and frequency resources with different power allocations. These lead to the higher spectral efficiency than the conventional orthogonal multiple access (OMA) [1-2].

Usually, a far user (FU) at the cell edge receives weak signal from a base station compared to the near users (NUs) in general cellular environments. This makes the more power allocation to FU to compensate the signal strength caused by the path loss and fading than NU in NOMA systems. Recently a cooperative communication with a relay is introduced in NOMA system to improve the performance degradation using a relay and/or space diversity against the fading in wireless channel. There are two kinds of relaying method which are the dedicated relaying [3-4], and the user relaying [5-6]. Firstly, the dedicated relaying is the conventional relaying method which receives signals from a base station and transmits to users by amplify-and-forward or decode-and-forward relaying. Secondly, the user relaying utilizes an NU in NOMA instead of a standalone dedicated relay. In the process of the successive interference cancellation (SIC) which removes interferences from other users in NOMA system, the decoded information of FU is obtained and it can be used for the relaying [5]. In the case of a temporary network, i.e., an Ad-Hoc network, the user relaying is more preferable than the permanent dedicated relaying for simple and temporary network configuration.

Furthermore, we can apply the relay selection which improves the performance of a system when multiple candidate relays are exist. In [4, 7], the performance improvements in a cooperative NOMA system with relay selection under the assumption of perfect channel state information (CSI) have been

International Journal of Intelligent Engineering and Systems, Vol.11, No.6, 2018

DOI: 10.22266/ijies2018.1231.15

analyzed. However, this assumption is not realistic in mobile fading environments. It is important to notice the time differences between the moments of the selection and the actual information relav transmission can cause channel errors in wireless fading channels. Consequently, these channel errors are seriously diminishing the performance gain obtained from the relay selection. Recently published studies of [8-10] consider the degradation incurred from the channel errors in NOMA system, but didn't expand to a cooperative system. In [9] and [10], the approximation and upper bound of the outage probability are derived, respectively. More recently, [11] discussed a cooperative NOMA with imperfect CSI, however, it considered the case of the dedicated relaying and did not include the relay selection.

Recently, we published user relay cooperative NOMA systems [12]. However, it assumed perfect channel state information, which does not include channel errors. Therefore, we consider a user relay cooperative NOMA system with channel errors. The max-min relay selection strategy is applied for the selection of a relay user (RU) among NUs in NOMA system. Also, the selection combining (SC) is adapted to FU for space diversity gain. We derive the outage probability of RU and FU in closed-form, respectively, since the information for RU and FU is different in NOMA systems. We showed the channel errors cause the sensitive degradation to the outage probability of RU and FU. To demonstrate the accuracy of the developed analytical results, we showed the simulation results perfectly match the analytical results.

The remainder of this paper is organized as follows. In section 2, the system model, channel errors, and relay selection strategy are described. Section 3 derives the outage probability of RU, FU, and the end-to-end performance in closed-form. The numerical results are presented and Monte Carlo simulations are applied to verify the accuracy of the derived analysis in section 4. Finally, section 5 concludes the paper.

2. System model

We consider a downlink NOMA system which consists of a base station as a source (S), N Relays (R_i , i = 1,2,..., N), and FU as shown in Fig.1. In this paper, we consider the user relaying, therefore N relays which are the NUs in the NOMA system are the candidate relays. We assume NUs are clustered relatively close together, hence the distances between S and NUs and between NUs and FU are identical, respectively [3].

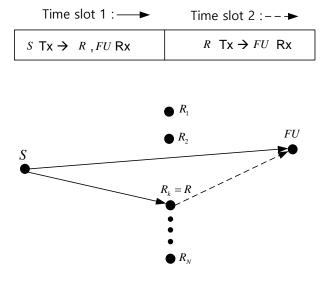


Figure. 1 System model

As mentioned in section 1, the selected relay removes the interferences from FU with the SIC process. The decoded information for FU during the process can be utilized for relaying. In Fig.1, the selected relay of R_k , which is the best relay to maximize the performance of FU, forwards the decoded information to FU. The max-min selection strategy is applied to select the best relay among NUs [13, 14]. We assume the best relay is selected prior to information transmission for multiplexing R_k and FU with different power levels in NOMA scheme.

The rapidly changing channel causes channel estimation errors. There are many factors to cause the channel estimation errors, feedback delay, fading and quantization errors. Statistically, the actual channel and the estimated channel can be represented by [15-16]

$$h = \rho \hat{h} + \sqrt{1 - \rho^2} \varepsilon \tag{1}$$

where *h* denotes actual channel coefficient and assumed as a complex Gaussian random variable, $h \sim CN(0,1)$. And the estimated channel coefficient is denoted by $\hat{h} \sim CN(0,1)$. ρ is the correlation coefficient between the actual and the estimated channel coefficients. $\varepsilon = CN(0,1)$ is independent of \hat{h} . All channels are assumed independent and identically distributed (i.i.d) Rayleigh block-fading channels; the channel response remains constant during one timeslot, but changes independently from one timeslot to another. We denote the actual channel coefficient and its estimated channel coefficient between node A and B as h_{AB} and \hat{h}_{AB} , respectively. The max-min relay selection strategy which is known as a near optimal selection method to solve the bottleneck phenomenon can be described as follows; a central controller which knows every CSIs through pilot channels selects the less channel gain link of $S - R_i$ or $R_i - FU$ link in $S - R_i - FU$ path. Next, select maximum channel gain path $S - R_i - FU$ among the selected N links (i.e., $S - R_i$ or $R_i - FU$ link, i = 1,2,..., N) [4], [14]. The index of the selected best relay at the relay selection stage can be written by

$$k = \arg \max_{i=1,2,\dots,N} \left(\left| \hat{h}_{SR_i} \right|^2, \left| \hat{h}_{R_iF} \right|^2 \right).$$
(2)

For convenience, denote the selected relay R_k by R. As shown in Fig.1, it requires two-time slots for the information transmission.

Time slot 1: Source transmit phase

At the first timeslot, a base station S transmits the multiplexed message to the selected relay R and FU based on the NOMA scheme, which is written by

$$s = \sqrt{P_S} \left(\sqrt{\alpha_R} x_R + \sqrt{\alpha_F} x_F \right)$$
(3)

where P_S is the transmit power of the base station. α_R and α_F are the power allocation coefficient for R and FU, respectively, where $\alpha_F > \alpha_R$ with $\alpha_F + \alpha_R = 1$. x_R and x_F denote the bipolar message symbol for R and FU, respectively, with $|x_R|^2 = |x_F|^2 = 1$.

The received signal of *R* from S - R path can be written by

$$y_{SR} = \sqrt{P_{SR}} \left(\sqrt{\alpha_R} x_R + \sqrt{\alpha_F} x_F \right) h_{SR} + n_R \tag{4}$$

where P_{SR} denotes average received power at R, which can be defined by $P_{SR} = P_S d_{SR}^{-\alpha}$ where d_{SR} is the distances between S and R, and α is the propagation loss constant. n_R denotes the Gaussian noise of R, $n \sim CN(0, N_0)$.

The received signal-to-interference plus noise ratio (SINR) of R to detect x_F can be written by

$$\gamma_{SR}^{X_F} = \frac{\alpha_F \left| h_{SR} \right|^2}{\alpha_R \left| h_{SR} \right|^2 + 1/\rho_{SR}}$$
(5)

where ρ_{SR} is the received signal-to-noise ratio (SNR) of *R*, $\rho_{SR} = P_{SR} / N_0$. After decoding x_F , this component is subtracted from the received signal by SIC. Thus the received instantaneous SNR of *R* to detect x_R after SIC can be written by

$$\gamma_{SR}^{X_R} = \alpha_R \,\rho_{SR} \left| h_{SR} \right|^2. \tag{6}$$

The received signal of FU from S - FU path can be written by

$$y_{SF} = \sqrt{P_{SF}} \left(\sqrt{\alpha_R} x_R + \sqrt{\alpha_F} x_F \right) h_{SF} + n_F$$
(7)

where P_{SF} denotes the average received power of FU. n_F denotes the Gaussian noise of the far user, $n_F \sim CN(0, N_0)$. Here we assumed the noise power of *R* and FU is identical.

The received SINR of FU from Eq. (7) can be given by

$$\gamma_{SF}^{x_{F}} = \frac{\alpha_{F} \left| h_{SF} \right|^{2}}{\alpha_{R} \left| h_{SF} \right|^{2} + 1/\rho_{SF}}$$
(8)

where ρ_{SF} is the received SNR of FU from S - FUpath, $\rho_{SF} = P_{SF} / N_0$.

Time slot 2: Relaying and combining phase

At this time slot, the selected decode-and-forward relay (R) transmits to FU. The received signal of FU from R - FU path can be written by

$$y_{RF} = \sqrt{\alpha_F P_{RF}} h_{RF} x_F + n_F \tag{9}$$

where P_{RF} denotes average received power at FU.

The received SNR is given by

$$\gamma_{RF}^{X_F} = \alpha_F \,\rho_{RF} \left| h_{RF} \right|^2 \tag{10}$$

where ρ_{RF} is the received SNR of FU from R - FUpath, $\rho_{RF} = P_{RF} / N_0$.

The received signals from both the indirect (R - FU) path and the direct (S - FU) path are combined by SC at FU.

3. Outage probability

The information for R is different from FU in NOMA, the outage probability of R and FU is derived respectively. Also the system outage probability of R and FU paired is derived.

3.1 Outage probability of R

The outage of the selected relay is happened in two cases. First, SINR of the far user information of x_D bellows the predefined threshold, hence x_D cannot be decoded. On the other hand, SNR of the selected relay after SIC bellows threshold, hence x_R cannot be decoded. Therefore, the outage probability of the selected relay can be written by

$$P_{o,R} = \Pr\left(\gamma_{SR}^{x_F} < \Gamma_F\right) + \Pr\left(\gamma_{SR}^{x_F} \ge \Gamma_F, \gamma_{SR}^{x_R} < \Gamma_R\right)$$
(11)

where Γ_F and Γ_R , $\Gamma_F = 2^{2R_F} - 1$ and $\Gamma_R = 2^{R_R} - 1$, are the threshold SNR. R_F and R_R are the spectral efficiency of FU and R, respectively.

Replacing Eq. (5) into the first probability of Eq. (11), we can write

$$\Pr\left(\gamma_{SR}^{X_{F}} < \Gamma_{F}\right) = \Pr\left(\frac{\alpha_{F} \left|h_{SR}\right|^{2}}{\alpha_{R} \left|h_{SR}\right|^{2} + 1/\rho_{SR}} < \Gamma_{F}\right)$$
(12)
$$= \Pr\left(\left|h_{SR}\right|^{2} < \varsigma\right)$$

where $\zeta = \Gamma_F / (\alpha_F - \Gamma_F \alpha_R) \rho_{SR}$, and $\Gamma_F < \alpha_F / \alpha_R$. And a part of the second probabilities,

$$\Pr\left(\gamma_{SR}^{x_{R}} < \Gamma_{R}\right) = \Pr\left(\alpha_{R} \rho_{SR} \left| h_{SR} \right|^{2} < \Gamma_{R}\right)$$

$$= \Pr\left(\left| h_{SR} \right|^{2} < \eta\right)$$
(13)

where $\eta = \Gamma_R / \alpha_R \rho_{SR}$.

Consequently, the outage probability of R from Eq. (12) and Eq. (13) can be written by

$$P_{o,R} = \begin{cases} \Pr\left(\left|h_{SR}\right|^{2} < \varsigma\right), \varsigma \ge \eta \\ \Pr\left(\left|h_{SR}\right|^{2} < \eta\right), \varsigma < \eta \end{cases}$$

$$= F_{H_{R}} \left\{\max\left(\varsigma, \eta\right)\right\}$$
(14)

where $|h_{SR}|^2$ is replaced to H_R , $|h_{SR}|^2 = H_R$, for notational simplicity in the last equality. $F_{H_R} \{\max(\varsigma, \eta)\}$ denotes the cumulative distribution function(CDF) of H_R . We will derive the probability distribution function (pdf) of $f_{H_R}(\bullet)$, firstly. Then, the CDF of $F_{H_R}(\bullet)$ can be obtained from $f_{H_R}(\bullet)$ by taking integration.

On the other hand, the actual channel and the estimated channel is different as mentioned in Eq. (1). Therefore the pdf of H_R can be obtained from the conditional pdf and written by

$$f_{H_{R}}(z) = \int_{0}^{\infty} f_{H_{R}} |\hat{H}_{R}(z|x) f_{\hat{H}_{R}}(x) dx$$
(15)

where $f_{H_R|\hat{H}_R}(z|x)$ denotes the conditional pdf with channel errors, and given by [17-18]

$$f_{H_R|\hat{H}_R}(z|x) = \frac{1}{1-\rho^2} e^{-\frac{Z+\rho^2 x}{1-\rho^2}} I_o\left(\frac{2\rho\sqrt{zx}}{1-\rho^2}\right).$$
(16)

Also the pdf of $\hat{H_R}$, $f_{\hat{H_R}}(x)$, denotes the pdf of the selected max-min relay in Eq. (15). From Eq. (23) in [12] which can be obtained from Eq. (2) in [14], it is given by

$$f_{\hat{H}_{R}}(x) = \sum_{i=1}^{N} \binom{N}{i} (-1)^{i-1} \frac{i}{1-2i} \left(e^{-2ix} - e^{-x} \right) + \sum_{i=1}^{N} \binom{N}{i} (-1)^{i-1} i e^{-2ix}$$
(17)

Replacing Eqs. (16) and (17) into Eq. (15), we can obtain

$$f_{H_{R}}(z) = \frac{1}{1-\rho^{2}} e^{-z/(1-\rho^{2})} \sum_{i=1}^{N} {\binom{N}{i}} (-1)^{i-1} \frac{1}{1-2i} \int_{0}^{\infty} e^{-\left(\frac{\rho^{2}}{1-\rho^{2}}+2i\right)^{k}} I_{0}\left(\frac{2\rho\sqrt{zx}}{1-\rho^{2}}\right) dx$$

$$-\frac{1}{1-\rho^{2}} e^{-z/(1-\rho^{2})} \sum_{i=1}^{N} {\binom{N}{i}} (-1)^{i-1} \frac{1}{1-2i} \int_{0}^{\infty} e^{-\left(\frac{\rho^{2}}{1-\rho^{2}}+1\right)^{k}} I_{0}\left(\frac{2\rho\sqrt{zx}}{1-\rho^{2}}\right) dx$$

$$+\frac{1}{1-\rho^{2}} e^{-z/(1-\rho^{2})} \sum_{i=1}^{N} {\binom{N}{i}} (-1)^{i-1} i \int_{0}^{\infty} e^{-\left(\frac{\rho^{2}}{1-\rho^{2}}+2i\right)^{k}} I_{0}\left(\frac{2\rho\sqrt{zx}}{1-\rho^{2}}\right) dx$$

(18)

From (6.614.3) and (9.220.2) in [19],

International Journal of Intelligent Engineering and Systems, Vol.11, No.6, 2018

$$\int_{0}^{\infty} e^{-\alpha x} I_{2\nu} \left(2\sqrt{\beta x} \right) dx = \frac{1}{\sqrt{\alpha \beta} \Gamma(2\nu+1)} e^{\frac{\beta}{2\alpha}} \Gamma(\nu+1)$$
$$\times M_{-1/2,\nu} \left(\frac{\beta}{\alpha}\right), \ \alpha > 0, \nu > -1$$
(19)

where

$$M_{\lambda,\mu}(z) = z^{\mu+1/2} e^{-z/2} \Phi\left(\mu - \lambda + \frac{1}{2}, 2\mu + 1; z\right)$$
(20)

and where $\Phi(\alpha, \alpha, z) = e^z$ [19, (9.215.1)]. Appling Eqs. (19) and (20) to Eq. (18), we can rearrange the pdf in closed-form

$$f_{H_{R}}(z) = \sum_{i=1}^{N} \binom{N}{i} (-1)^{i-1} \frac{i}{1-2i} \frac{1}{\rho^{2}+2(1-\rho^{2})} e^{-\frac{z}{1-\rho^{2}} \left\{ 1 - \frac{\rho^{2}}{\rho^{2}+2(1-\rho^{2})i} \right\}} \\ - \sum_{i=1}^{N} \binom{N}{i} (-1)^{i-1} \frac{i}{1-2i} e^{-z} \\ + \sum_{i=1}^{N} \binom{N}{i} (-1)^{i-1} \frac{i}{\rho^{2}+2(1-\rho^{2})i} e^{-\frac{z}{1-\rho^{2}} \left\{ 1 - \frac{\rho^{2}}{\rho^{2}+2(1-\rho^{2})i} \right\}}$$
(21)

By integrating Eq. (21), the CDF of H_R can be written by

$$F_{H_{R}}(y) = \int_{0}^{y} f_{H_{R}}(z)dz$$

= $\sum_{i=1}^{N} {N \choose i} (-1)^{i-1} \frac{i}{1-2i} \frac{1}{\rho^{2}+2(1-\rho^{2})i} \Phi_{a}(y)$ (22)
 $-\sum_{i=1}^{N} {N \choose i} (-1)^{i-1} \frac{i}{1-2i} (1-e^{-y})$
 $+ \sum_{i=1}^{N} {N \choose i} (-1)^{i-1} \frac{i}{\rho^{2}+2(1-\rho^{2})i} \Phi_{a}(y)$

where

and

$$a = \frac{1}{1 - \rho^2} \left\{ 1 - \frac{\rho^2}{\rho^2 + 2(1 - \rho^2) i} \right\}.$$

 $\Phi_a(y) = \frac{1}{1 - e^{-ay}}$

Therefore, $F_{H_R} \{\max(\varsigma, \eta)\}$ can be obtained from substituting γ by max (ς, η) in Eq. (22). As a special case of $\rho = 1$, which means error free, Eq. (22) becomes identical to [12, (23)].

3.2 Outage probability of FU

The received signals both from the indirect path (R - FU path) and from the direct path (S - FU path)

path) are combined at FU with SC. Therefore, the outage probability of FU can be obtained by multiplying the outage probability of the indirect path and that of the direct path, and can be written by

$$P_{o,F} = P_{o,ind} \times P_{o,dir} \tag{23}$$

where $P_{o,ind}$ and $P_{o,dir}$ denote the outage probability of the indirect path and the direct path, respectively. The indirect path are composed of S - R and R - FU links. When one of the link fails, the outage happens. The outage probability of FU through indirect path can be written by

$$P_{o,ind} = \Pr\left\{ \min\left(\gamma_{SR}^{x_{F}}, \gamma_{RF}^{x_{F}}\right) < \Gamma_{F} \right\}$$

$$= 1 - \Pr\left(\gamma_{SR}^{x_{F}} \ge \Gamma_{F}\right) \Pr\left(\gamma_{RF}^{x_{F}} \ge \Gamma_{F}\right)$$

$$= 1 - \Pr\left(H_{R} \ge \varsigma\right) \Pr\left(H_{RF} \ge \lambda\right)$$

$$= 1 - \left\{1 - F_{H_{R}}\left(\varsigma\right)\right\} \left\{1 - F_{H_{RF}}\left(\lambda\right)\right\}$$

$$(24)$$

where $H_{RF} = |h_{RF}|^2$ and $\lambda = \Gamma_F / \rho_{RF} \alpha_F$. Similarly, the pdf of H_{RF} can be obtained from Eq. (15) by replacing H_{RF} and \hat{H}_{RF} instead of H_R and \hat{H}_R , respectively. We can write

$$f_{H_{RF}}(z) = \int_{0}^{\infty} f_{H_{RF}|\hat{H}_{RF}}(z|x) f_{\hat{H}_{RF}}(x) dx.$$
 (25)

Also the CDF of $F_{H_{RF}}(\lambda)$ can be obtained by substituting H_{RF} and λ by H_R and y in Eq. (22), respectively.

On the other hand, the outage probability of FU through direct path can be derived similar to Eq. (12), and can be written by

$$P_{o,dir} = \Pr\left(\gamma_{SF}^{x_F} < \Gamma_F\right) = \Pr\left(\left|h_{SF}\right|^2 < \chi\right) = 1 - e^{-x} \quad (26)$$

where $\chi = \Gamma_F / \rho_{SF} (\alpha_F - \alpha_R \Gamma_F)$, $\rho_{SF} = P_{SF} / N_0$, and $P_{SF} = P_S d_{SF}^{-\alpha}$ which is the average received power from direct path. The last equality of Eq. (26) assumed Rayleigh fading.

Finally, the outage probability of FU with SC can be obtained by replacing Eqs. (24) and (26) into Eq. (23).

International Journal of Intelligent Engineering and Systems, Vol.11, No.6, 2018

3.3 Outage probability of NOMA system

The outage of NOMA system happens either R or FU is in outage. Hence the end-to-end outage probability of NOMA system can be defined by

$$P_{o,sys} = 1 - \left(1 - P_{o,R}\right) \left(1 - P_{o,F}\right). \tag{27}$$

By inserting Eqs. (14) and (23) into Eq. (27), we can written

$$P_{o,sys} = 1 - \left[1 - F_{H_R} \left\{ \max\left(\varsigma, \eta\right) \right\} \right] \times \left[1 - \left\{F_{H_R}\left(\varsigma\right) + F_{H_{RF}}\left(\lambda\right) - F_{H_R}\left(\varsigma\right)F_{H_{RF}}\left(\lambda\right)\right\} + \left[1 - e^{-\chi}\right]\right].$$
(28)

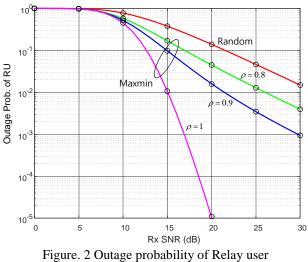
4. Numerical examples

Fig. 2 shows the outage probability of the selected relay user R, where the solid lines denote the analytical results and "o" indicates the Monte Carlo simulation results. The analytical results and the simulation results are perfectly matched.

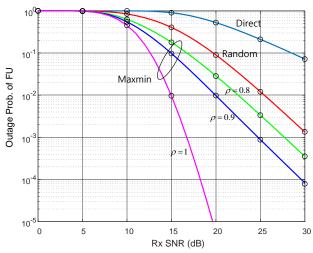
In Fig. 2, the curb "Random" is also plotted for reference, where "Random" means that the relay is selected randomly among NUs. While "Maxmin" denotes the relay with the max-min selection strategy. Especially, the outage probability with $\rho = 1$ displays the special case of no channel error, which coincides with our previous results in [12]. The performance improvements with the max-min relay selection decrease as the value of ρ decreases. When $\rho = 0$, the outage probability coincides with that of the random selection. However, we noticed that the performance improvement with the max-min relay selection is highly effective at low channel error, that means high correlation coefficient ρ .

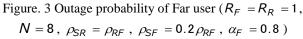
The outage probability of FU versus average SNR shows in Fig. 3. In this figure, the curb "Direct" is also plotted for reference, where "Direct" means the outage probability of FU with the direct path (S - FU path) only. Irrespective of relay selection method, max-min or random relay selection, we noticed that the performance of the cooperative system is better than that of the direct path only.

Similar to the results of Fig. 2, the performance with the max-min relay selection improves as ρ approaches to 1. On the contrary, the performance degrades as ρ approaches to 0, and finally identical to that of the random selection. It is noticed that the effect of the channel error to the performance of FU is less sensitive compared to that of RU, since the space diversity of SC compensates the degradation.



 $(R_F = R_R = 1, N = 8, \rho_{SR} = \rho_{RF}, \alpha_F = 0.8)$





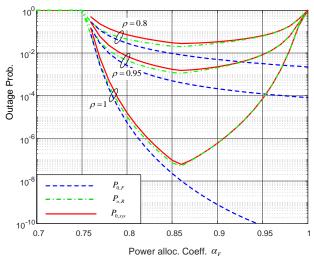


Figure. 4 Outage probability vs. power allocation for different correlation coefficients ($R_F = R_R = 1, N = 8$, $\rho_{SR} = \rho_{RF} = 20 dB$, $\rho_{SF} = 0.2 \rho_{RF}$)

The effect of the power allocation to the outage probabilities are shown in Fig. 4. The increase of α_F means the power allocation to FU increases in Eq. (3), consequently the outage probability of FU decreases. However the total transmit power is limited to $\alpha_F + \alpha_R = 1$, it is noticed that the optimal power allocation to minimize the outage probability of *R* exists at the given condition as shown in Fig.4. Also we noticed that the system outage probability dominantly depends on the outage probability of relay *R*.

This Fig.4 shows the power allocation coefficient to minimize the outage probability under given conditions is approximately 0.86 irrespective of the channel correlation coefficient. As the channel correlation coefficient decrease, the effect of power allocation to the performance becomes lessen.

5. Conclusions

In this paper, we consider a user relay cooperative NOMA system with channel estimation errors. The channel estimation errors are modeled by a correlation coefficient between the actual channel and the estimated channel. The max-min relay selection strategy is applied for the relay selection among NUs and the selection combining is assumed for the space diversity to FU. The outage performances of RU and FU as a function of the correlation coefficient are derived analytically in closed-form, respectively. The derived performances are verified by Monte Carlo simulation.

As the previous studies found, the max-min relay selection without channel errors ($\rho = 1$) is effective to improve the outage performance compared to the random selection which is shown in Fig.2 and Fig.3. However, these figures demonstrate that the outage probabilities of RU and FU are affected by the channel correlation. In Fig.2 and Fig.3, to maintain the outage probability of 1×10^{-3} with $\rho = 0.9$, the relay user and the far user require more than 13 dB and 8 dB of SNR compared with $\rho = 1$, respectively. It represents that the outage probability of the relay user is more susceptible to the channel errors. It is interpreted that the spatial diversity at far user which combine both signals from the direct and indirect paths mitigates the performance degradation caused by channel errors.

Moreover, the channel correlation coefficient approaches to 0, the performance gains are banishing and the outage probability with the max-min relay selection becomes identical to that with the random relay selection. From this result, we can conclude that the effect of the channel errors to the outage performance is not negligible. The generally accepted performance gain from the max-min relay selection can be banished as the channel errors increasing.

NOMA system has an optimal power allocation coefficient to minimize the outage performance, but as the correlation coefficient decrease, the effect of power allocation to the performance becomes lessen. Further research will be focused on the performance of cooperative NOMA systems in general fading.

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International Journal of Intelligent Engineering and Systems, Vol.11, No.6, 2018

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